

Applicant : Gunther Bergk  
Serial No. : Not Yet Assigned  
Filed : Herewith  
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Amendments to the Drawings:

The attached replacement drawing sheet includes changes to the figure and replaces the original sheet.

The figure was amended to include a figure reference (i.e., "FIG. 1").

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REMARKS

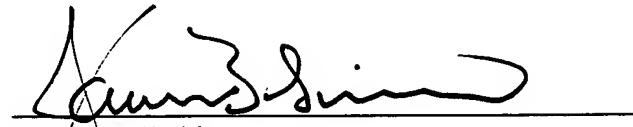
It should be noted that the edits in the substitute specification primarily represent a reorganization of the subject matter of the original priority application and do not represent new matter. Attached to substitute specification are clean versions of the amended claims and abstract, for ease of publication.

In view of the above, Applicant submits that the specification and all claims meet the formal requirements for issuance, and respectfully request allowance.

Please apply any charges or credits to deposit account 06-1050, referencing attorney docket number 02894-734US1.

Respectfully submitted,

Date: January 31, 2006

  
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## ~~CIRCUIT ARRANGEMENT FOR THE INDUCTIVE TRANSMISSION OF ELECTRIC ENERGY~~

### TECHNICAL FIELD

~~The invention pertains~~ This disclosure relates to a ~~circuit arrangement for the~~ inductive transmission of electric energy, for example, a circuit arrangement for supplying an ~~small~~ electrical appliance with power and/or for inductively charging a battery.

### BACKGROUND

Devices for transmitting electric energy, such as electrical circuits, are known in the art. For example, DE 38 42 465 A1 discloses a switching controller for a d.c.-d.c. conversion which comprises an electronic switch and a series-resonant circuit (rather than instead of an inductor). The switching controller oscillates with the resonant frequency of the series-resonant circuit and therefore has a particularly high efficiency if the electronic switch switches in the zero crossings of the current. The electronic switch is realized with two complimentary switching transistors that are controlled in antiphase. The control of the switching transistors is realized with a feedback circuit and input stages for the switching transistors that are not described in detail.

As another example, DE 40 15 455 A1 discloses a control circuit for an inverted rectifier that comprises a push-pull output stage with complementary transistors. The control of the transistors is realized with two electrically coupled control signals of mutually shifted potential. The connecting and disconnecting control signal edges are shifted by means of a delay circuit such that the initially switched-on transistor is switched off before the still switched-off transistor is switched on. This results in a relatively complex means that this control circuit is relatively complex.

### SUMMARY

The present invention is based on the objective of disclosing a ~~circuit arrangement~~ for the inductive transmission of electric energy that makes it possible to achieve a high efficiency with a low circuit expenditure.

According to the invention one aspect, this objective is attained with an inductive electric energy transmission circuit arrangement with the following characteristics includes: an oscillating circuit; a push-pull circuit with including

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complimentary-a first switching transistors and a second switching transistor that are adapted configured to excite alternate a current flow through the oscillating circuit. The circuit arrangement also includes, a control circuit including for the complimentary switching transistors that preferably also contains complimentary first and second control transistors, configured to control the first and second switching transistors; and a frequency generator, the configured to generate an output signal adapted to drive of which can be fed to the control transistors. With a low circuit expenditure, a high efficiency circuit arrangement for transmission of electric energy can be made available.

With respect to a high efficiency of the circuit arrangement, it is advantageous to at least approximately tune. In some cases, the oscillating circuit is configured to oscillate at an oscillatory to the frequency substantially equal to a frequency of the output signal of the frequency generator.

In some cases, the output signal of the frequency generator includes. It is also advantageous with respect to a reduction of the power loss in the switching transistors if these transistors can be controlled with a square-wave signal, i.e., if the frequency generator delivers a square-wave output signal. In addition, the initially switched-on switching transistor should already be switched off before the still switched-on switching transistor is switched on in order to prevent the voltage source from being quasi short-circuited by the two switching transistors that are simultaneously switched on. This is the reason why the control circuit for the switching transistors comprises a resistor and two capacitors in addition to the complimentary control transistors in one preferred embodiment of the circuit arrangement according to the invention.

In some embodiments, a control terminal of the first control transistor and a control terminal of the second control transistor are configured to receive the output signal from the frequency generator. The resistor connects the. Preferably, a control terminals of the first switching transistor is electrically connected to a first end of a resistor, and a control terminal of the second switching transistor is connected to a second end of the resistor.

In some implementations, the a first capacitor is arranged electrically parallel to the a main current path of the first control transistor, wherein a first end of the capacitor is electrically connected to the first end of the resistor; and the a second capacitor is arranged electrically parallel to the a main current path of the second control transistor, wherein the first capacitor is connected to the first end of the resistor and the a first end of the second capacitor is electrically connected to the second end of the resistor.

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Preferably, the first capacitor, the resistor, and the second capacitor form a series connection, wherein a supply voltage source is connected in parallel with the series connection. Due to these measures, one switching transistor is can be switched off faster than the other switching transistor is switched on, thereby preventing the supply voltage source from being quasi short-circuited as a result of the first and second switching transistors being simultaneously switched on.

In some examples, The the oscillating circuit has includes an inductor-inductive coil and a capacitor. In this case, The the inductor-inductive coil can be realized, for example, in the form of a coil that may form the a primary coil of a transformer. A Preferably, the primary coil can supply electric energy to a secondary coil can deliver electric energy in this case of the transformer. The circuit arrangement according to the invention can be used, for example, for supplying electric energy to a small electrical appliance that contains the secondary coil, preferably electric toothbrushes or electric razors that may also contain a battery.

The invention is described below with reference to oneOne embodiment of a circuit arrangement according to the inventionconfigured for the inductive transmission of electric energy that is illustrated in the only figureFIG. 1. Other embodiments are discussed in the following description.

DESCRIPTION OF DRAWINGS

FIG. 1 is a schematic diagram of circuit arrangement configured to inductively transmit electric energy.

DETAILED DESCRIPTION

The circuit arrangement according to the invention that is illustrated in the figureFIG. 1 comprises includes two switching transistors in the form of complimentary field effect transistors (e.g., first switching transistor T2, and second switching transistor T4), the drain terminals of which the first switching transistor T2 and second switching transistor T4 are connected to one another and to one end of an resonant oscillating circuit consisting of including an inductor Lpr and a capacitor Cpr. The other end of the resonant oscillating circuit and the source terminal of the n-channel field effect transistor (i.e., second switching transistor T4) are connected to ground. The source terminal of the p-channel field effect transistor (i.e., first switching transistor T2) is connected to the plus pole of a supply voltage source U1. The minus pole of the supply voltage source U1 is connected to ground. The circuit arrangement comprises includes two additional control transistors in the form of complimentary field effect transistors (e.g., first control

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transistor T1, and second control transistor T3), the gate terminals of which the first control transistor T1 and the second control transistor T3 are directly connected to the input of a frequency generator F that generates an output signal referred to ground. The gate terminals of the first and second switching transistors T2, T4 are connected by means of a resistor R1. One end of the resistor R1 is connected to the drain terminal of the p-channel field effect transistor (i.e., the first control transistor T1), as well as to the plus pole of the supply voltage source U1 by means of the first capacitor C1. The other end of the resistor R1 is connected to the drain terminal of the n-channel field effect transistor (i.e., the second control transistor T3), as well as to ground by means of a second capacitor C2. The source terminal of the p-channel field effect first control transistor T1 is connected to the plus pole of the supply voltage source U1. The minus pole of the supply voltage source U1 is connected to the source terminal of the n-channel field effect second control transistor T3.

\_\_\_\_\_ In another embodiment of a circuit arrangement according to the invention, the complementary field effect transistors first and second switching transistor T2, T4 and/or the complementary field effect transistors first and second control transistors T1, T2-T3 are can be replaced with complimentary bipolar transistors.

\_\_\_\_\_ In another embodiment of a circuit arrangement according to the invention, the polarities of the supply voltage source and the transistors (i.e., the first and second switching transistors and the first and second control transistors) are reversed.

\_\_\_\_\_ Other variations of the above-described embodiments of a circuit arrangement according to the invention lack the first and second capacitors C1 and C2, i.e., the function of these capacitors is respectively fulfilled by the gate-source capacitance and the base-emitter capacitance of the first and second switching transistors T2, T4.

\_\_\_\_\_ The function of the circuit arrangement illustrated in the figure FIG. 1 is briefly described below. The first and second switching transistors T2, T4 are wired in the from form of a push-pull stage and alternately connect the resonant oscillating circuit to the operating voltage and to ground, wherein the push-pull stage is controlled with square-wave signals by the first and second control transistors T1, T3. The first and second capacitors C1, C2 and the resistor R1 are provided in order to prevent that the switching transistors T2, T4 from simultaneously carrying a high drain current. For example, if the output signal of the frequency generator F is positive, i.e., if it approximately assumes the operating voltage, the second control transistor T3 is conductive and the first control transistor T1 is non-conductive. Consequently, the first switching transistor T2 is

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conductive and the second switching transistor T4 is non-conductive. When the output signal of the frequency generator F changes to ground, i.e., the reference potential, the first control transistor T1 and the second switching transistor T4 become conductive while the second control transistor T3 and the first switching transistor T2 become non-conductive. This causes the gate-source voltage of the first switching transistor T2 to drop with a time constant  $R'C'$ , wherein R' refers to the track resistance of the now conductive first control transistor T1 and C' refers to the sum of the capacitance of the first capacitor C1 and the input capacitance of the first switching transistor T2. The gate-source voltage of the second switching transistor T4 simultaneously increases with a time constant  $R''C''$ , wherein R'' refers to the sum of the resistance of the resistor R1 and the track resistance of the now conductive first control transistor T1 and C'' refers to the sum of the capacitance of the second capacitor C2 and the input capacitance of the switching transistor T4. Assuming that C' is practically equal to C'',  $R'C'$  is much shorter than  $R''C''$  because the resistance R' is much lower than the resistance R'', i.e. the switching transistor T2 is switched off faster than the switching transistor T4 is switched on. If the first and second capacitors C1, C2, the input capacitances of the first and second switching transistors T2, T4 and the track resistances of the first and second control transistors T1, T3 are approximately equal, one switching transistor consequently is always switched off faster than the other switching transistor is switched on. The time delay between switching on and switching off can be adapted to the switching and delay times of the first and second switching transistors T2, T4 by choosing the ratings of the first and second capacitors C1, C2 and of the resistor R1 accordingly.

FIG.1

